

Oppgave 1:

a) Vi kaller de norm. vinkler for henholdsvis

$$\hat{\omega}_1 = \pi, \hat{\omega}_2 = 2\pi/3, \hat{\omega}_3 = \pi/3, \hat{\omega}_4 = \pi/90$$

vi; her

$$\hat{\omega} = \frac{2\pi f}{f_s} \Rightarrow f = \frac{\hat{\omega} \cdot f_s}{2\pi}$$

så vi får (for $f_s = 900 \text{ Hz}$)

$$f_1 = f_s/2 = 450 \text{ Hz}$$

$$f_2 = f_s/3 = 300 \text{ Hz}$$

$$f_3 = f_s/6 = 150 \text{ Hz}$$

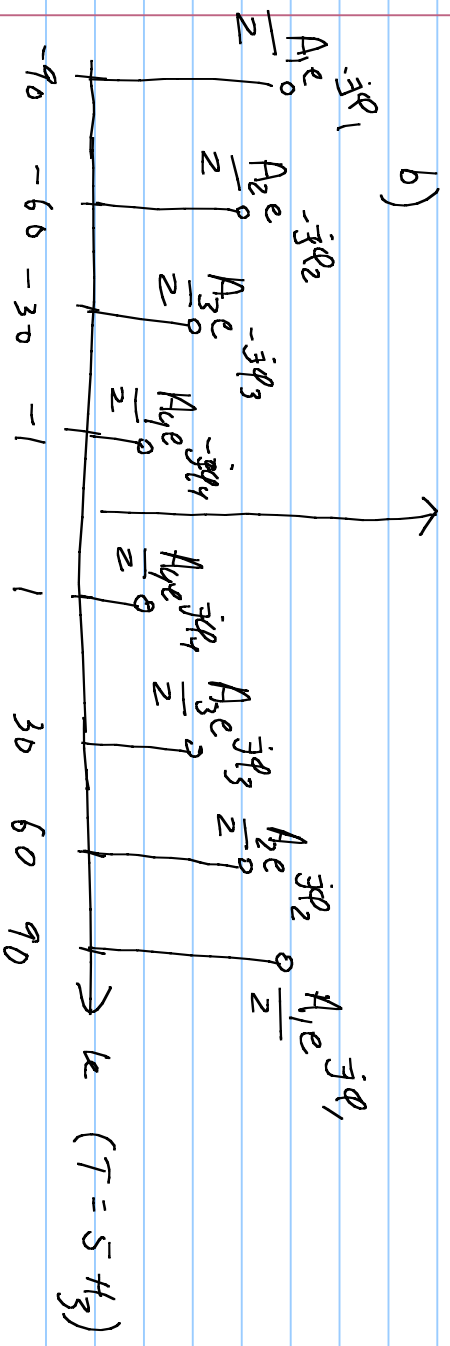
$$f_4 = f_s/180 = 5 \text{ Hz}$$

Så vi gir:

$$x(t) = A_1 \cos(900\pi t + \varphi_1) + A_2 \cos(600\pi t + \varphi_2) + A_3 \cos(300\pi t + \varphi_3) \\ + A_4 \cos(10\pi t + \varphi_4)$$

Signalet er periodisk fordi

$$\text{gcd}(450, 300, 150, 5) = 5$$



Oppgave 2:

$$\begin{aligned} c) \quad y[n] &= |H(e^{j\pi})| A_1 \cos(\pi n + \varphi_1) + \mathcal{R}\{H(e^{j\pi/3})\} \\ &\quad + |H(e^{j2\pi/3})| A_2 \cos\left(\frac{2\pi}{3}n + \varphi_2\right) + \mathcal{R}\{H(e^{j2\pi/3})\} \\ &\quad + |H(e^{j\pi/3})| A_3 \cos\left(\frac{\pi}{3}n + \varphi_3\right) + \mathcal{R}\{H(e^{j\pi/3})\} \\ &\quad + |H(e^{j\pi/90})| A_4 \cos\left(\frac{\pi}{90}n + \varphi_4\right) + \mathcal{R}\{H(e^{j\pi/90})\} \end{aligned}$$

\mathcal{R} : reell del (i det minste):

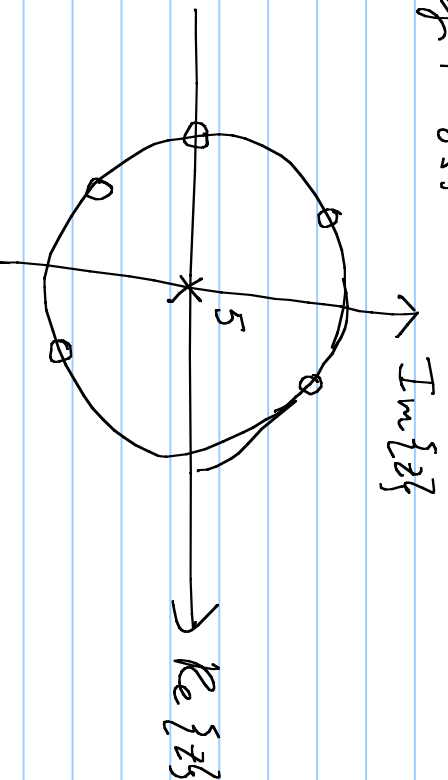
$$|H(e^{j\pi})| = |H(e^{j2\pi/3})| = |H(e^{j\pi/3})| = 0$$

$$|H(e^{j\pi/90})| \neq 0$$

$$\begin{aligned} b) \quad H(z) &= 0 \quad \text{for } z = e^{j\pi} = -1 \\ &= e^{j2\pi/3} \\ &= e^{j\pi/3} \end{aligned}$$

$$= e^{j\pi/3}$$

gitt oss



NB: Kravet om kausalitet gjør at vi har 5 poler i $z=0$!

c) Vi vet hvor røttene til polynomet $H(z)$ er, så vi kan skrive det på faktoriseret form:

$$H(z) = (1 - z^{-1} e^{j\pi/3}) (1 - z^{-1} e^{j2\pi/3}) (1 - z^{-1} e^{-j\pi/3}) (1 - z^{-1} e^{-j2\pi/3})$$

Så kan vi skrive det til

$$\begin{aligned} H(z) &= (1 - z^{-1} e^{j\pi/3}) (1 - 2 \cos(\frac{2\pi}{3}) z^{-1} + z^{-2}) (1 - 2 \cos(\frac{\pi}{3}) z^{-1} + z^{-2}) \\ &= (1 + z^{-1}) (1 + z^{-1} + z^{-2}) (1 - z^{-1} + z^{-2}) \\ &= (1 + z^{-1}) (1 + z^{-2} + z^{-4}) \\ &= (1 + z^{-2} + z^{-4} + z^{-1} + z^{-3} + z^{-5}) \\ &= \underline{\underline{1 + z^{-1} + z^{-2} + z^{-3} + z^{-4} + z^{-5}}} \end{aligned}$$

Det er ikke den eneste gyldige systemformulering, fordi

fordi $AH(z) = 0 \Leftrightarrow H(z) = 0$ for $A \neq 0$ (en konstant).

$$d) h[n] = \sum_{k=0}^5 \delta[n-k]$$

Hvis vi velger $A = \frac{1}{6}$ i formik delopgave, får vi

$$h[n] = \frac{1}{6} \sum_{k=0}^5 \delta[n-k], \text{ dvs. glidende-middel}$$

filter med 6 koeffisienter.

Opfrage 3:

$$H(e^{j\hat{\omega}}) = \frac{1}{6} \sum_{n=0}^5 e^{-j\hat{\omega}n} = \frac{1 - e^{-j6\hat{\omega}}}{6(1 - e^{-j\hat{\omega}})}$$

$$= \frac{\sin(3\hat{\omega})}{6 \sin(\hat{\omega}/2)} e^{-j5\hat{\omega}/2}$$

1) hier $\sin(3\hat{\omega}) < 0$ für $\pi/3 < |\hat{\omega}| < \frac{2\pi}{3}$

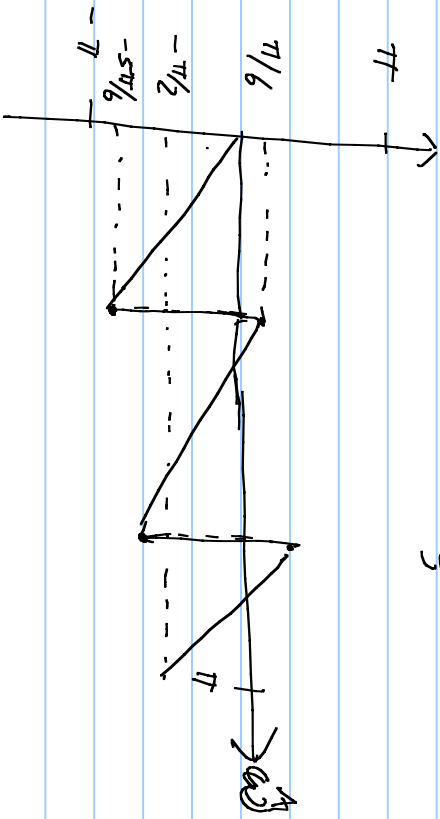
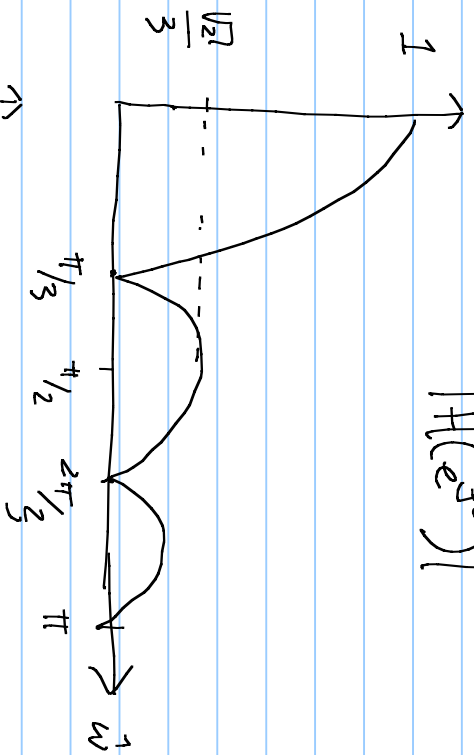
ist instabil für $-\pi < \hat{\omega} < \pi$, sei mir fair

$$|H(e^{j\hat{\omega}})| = \frac{|\sin(3\hat{\omega})|}{6|\sin(\hat{\omega}/2)|}$$

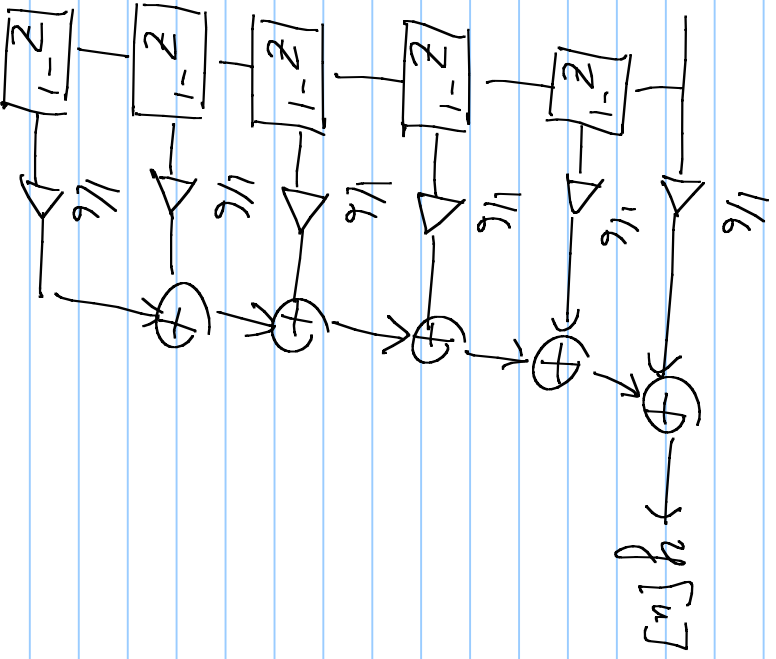
$$\neq H(e^{j\hat{\omega}}) = \begin{cases} -5\hat{\omega}/2 + \pi & \text{für } \pi/3 < |\hat{\omega}| < \frac{2\pi}{3} \\ -5\hat{\omega}/2 & \text{anders} \end{cases}$$

$$\text{oder } H(e^{j\hat{\omega}}) = |H(e^{j\hat{\omega}})| e^{j\phi H(e^{j\hat{\omega}})}$$

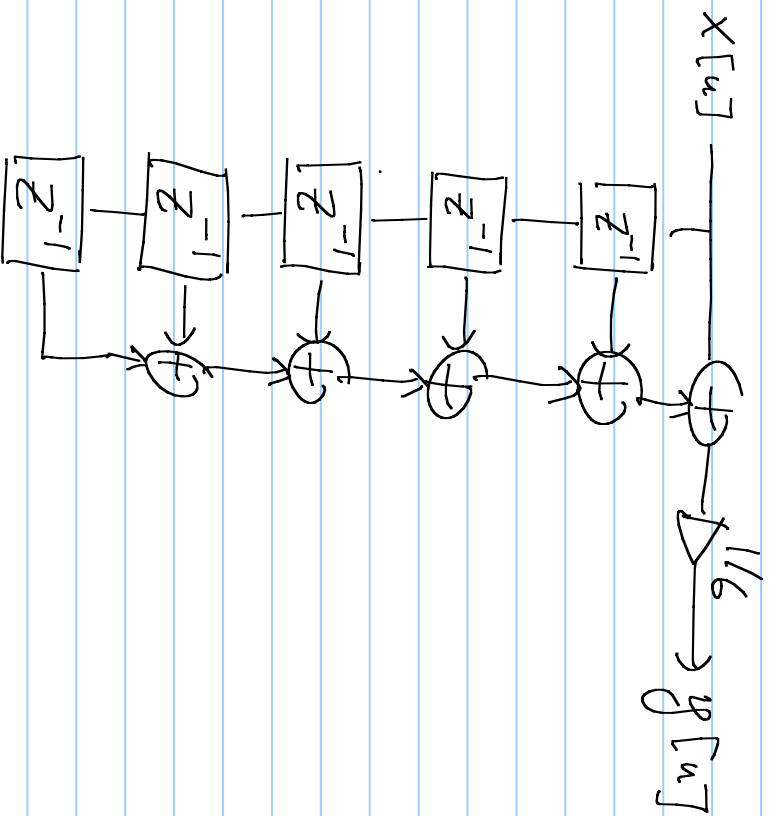
b) $|H(e^{j\omega})|$



c) $x[n] \rightarrow y[n]$



Kan skrives enklere:



Oppgave 4: $\tau+T$

$$K_u = \frac{1}{T} \int_{\tau}^{\tau+T} y(t) e^{-j2\pi kt/T} dt$$

$$= \frac{1}{T} \int_{\tau+T}^{\tau} x(-t) e^{-j2\pi kt/T} dt$$

$$t = \tau \Rightarrow \theta = -\tau$$

$$t = \tau+T \Rightarrow \theta = -\tau-T$$

$$\frac{d\theta}{dt} = -1 \Rightarrow d\theta = -dt$$

vi får $-T-T$

$$b_k = \frac{1}{T} \int_{-T}^{-T-T} x(\theta) e^{j2\pi k \theta / T} d\theta$$

$$= \frac{1}{T} \int_{-T-T}^{-T} x(\theta) e^{j2\pi k \theta / T} d\theta$$

$$= \left(\frac{1}{T} \int_{-T-T}^{-T} x(\theta) e^{-j2\pi k \theta / T} d\theta \right)^*$$

$$= \underline{\underline{(a_k)^* = a_{-k}}}$$

Så: $\boxed{b_k = a_{-k}}$

Forklaring: Et periodiske signal $x(t)$ kan skrives ud ved brug af formelen som en sum af sinusoider med givet amplitude og fasebølg. Hvert enkelt signal betyder at hver enkelt har en visse sinusoidalitet. Hvert enkelt sinusoidalitet betyder at hver enkelt har en visse sinusoidalitet. Hvert enkelt sinusoidalitet betyder at hver enkelt har en visse sinusoidalitet.

Derfor er det naturligt at Fourier-koefficienterne til $x(-t)$ er lige dem som tilhører $x(t)$ med modsat fortegn på fasen!

Oppgave 5:

- a) $h[n] \in \mathbb{R}$ $\forall n$ gir oss allm. linj. imp.
i) $H(e^{j\hat{\omega}})$ for $0 \leq \hat{\omega} \leq \pi$.

- b) Vi har $H(e^{j\hat{\omega}}) = H^*(e^{-j\hat{\omega}})$, så

$$|H(e^{j\hat{\omega}})| \text{ for } -\pi < \hat{\omega} < 0 = |H(e^{j\hat{\omega}})| \text{ for } 0 < \hat{\omega} < \pi$$

og $\angle H(e^{j\hat{\omega}})$ for $-\pi < \hat{\omega} < 0 = -\angle H(e^{j\hat{\omega}})$ for $0 < \hat{\omega} < \pi$

Oppgave 6:

$$\begin{aligned} \text{i) } Y_1^T \alpha x_1[n] + \beta x_2[n] &= (n+1) (\alpha x_1[n] + \beta x_2[n]) \\ &+ \alpha x_1[n-1] + \beta x_2[n-1] \\ \alpha Y_1^T x_1[n] + \beta Y_2^T x_2[n] &= \alpha (x_1[n] + x_1[n-1]) \\ &+ \beta ((n+1)x_2[n] + x_2[n-1]) \\ &= (n+1) (\alpha x_1[n] + \beta x_2[n]) \\ &+ \alpha x_1[n-1] + \beta x_2[n-1] \end{aligned}$$

Systemet er lineært.

$$y[n-k] = (n+1-k)X[n-k] + X[n-1-k]$$

$$T\{X[n-k]\} = (n+1)X[n-k] + X[n-1-k]$$

Systemet er ikke tidinvariant.

ii) Dette er et FIR-filter, som er LTI. (NB: Vi's helst ikke på vanlig måte.)

$$\text{iii) } T\{\alpha X_1[n] + \beta X_2[n]\} = \frac{\alpha X_1[n] + \beta X_2[n]}{\alpha X_1[n-1] + \beta X_2[n-1]}$$

$$\alpha T\{X_1[n]\} + \beta T\{X_2[n]\} = \alpha \frac{X_1[n]}{X_1[n-1]} + \beta \frac{X_2[n]}{X_2[n-1]}$$

$$= \frac{\alpha X_1[n] X_2[n-1] + \beta X_2[n] X_1[n-1]}{X_1[n-1] X_2[n-1]}$$

Systemet er ikke lineært

$$y[n-k] = \frac{X[n-k]}{X[n-k-1]}$$

$$T\{X[n-k]\} = \frac{X[n-k]}{X[n-k-1]} \left. \vphantom{T\{X[n-k]\}} \right\} \text{Systemet er tidinvariant}$$

